

Wing Connected Network Dual Voltage Source Inverter with the Ability to Improve the Quality Characteristics

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ABSTRACT

The main aim of project is to improve the power quality by using fuzzy based grid connected dual voltage source inverter. The proposed scheme is comprised of two inverters, which enables the microgrid to exchange power generated by the distributed energy resources (DERs) and also to compensate the local unbalanced and nonlinear load. The control algorithms are developed based on instantaneous symmetrical component theory (ISCT) to operate DVSI in grid sharing and grid injecting modes. The proposed scheme has increased reliability, lower bandwidth requirement of the main inverter, lower cost due to reduction in filter size, and better utilization of microgrid power while using reduced dc-link voltage rating for the main inverter. These features make the DVSI scheme a promising option for microgrid supplying sensitive loads. The proposed topology validated through simulation results.

Index Terms: Power quality, Distributed energy source, Fuzzy controller, Grid connected inverter, instantaneous symmetrical component theory (ISCT).

I. INTRODUCTION

The proliferation of power electronics devices and electrical loads with unbalanced nonlinear currents has degraded the power quality in the power distribution network. Moreover, if there is a considerable amount of feeder impedance in the distribution systems, the propagation of these harmonic currents distorts the voltage at the point of common coupling (PCC). At the same instant, industry automation has reached to a very high level of sophistication, where plants like automobile manufacturing units, chemical factories, and semiconductor industries require clean power.

For these applications, it is essential to compensate nonlinear and unbalanced load currents

. Load compensation and power injection using grid interactive inverters in microgrid have been presented in the literature. A single inverter system with power quality enhancement is discussed in. The main focus of this work is to realize dual functionalities in an inverter that would provide the active power injection from a solar PV system and also works as an active power filter, compensating unbalances and the reactive power required by other loads connected to the system. In , a voltage regulation and power flow control scheme for a wind energy system (WES) is proposed. A distribution static compensator (DSTATCOM) is utilized for voltage regulation and also for active power injection. The control scheme maintains the power balance at the grid terminal during the wind variations using sliding mode control. A multifunctional power electronic converter for the DG power system is described in .

This scheme has the capability to inject power generated by WES and also to perform as a harmonic compensator. Most of the reported literature in this area discuss the topologies and control algorithms to provide load compensation capability in the same inverter in addition to their active power injection. When a grid-connected inverter is used for active power injection as well as for load compensation, the inverter capacity that can be utilized for achieving the second objective is decided by the available instantaneous microgrid real power . Considering the case of a grid-connected PV inverter, the available capacity of the inverter to supply the reactive power becomes less during the maximum solar insolation periods.

At the same instant, the reactive power to regulate the PCC voltage is very much needed during this period. It indicates that providing multi functionalities in a single inverter degrades either

the real power injection or the load compensation capabilities. This paper demonstrates a dual voltage source inverter (DVSII) scheme, in which the power generated by the microgrid is injected as real power by the main voltage source inverter (MVSII) and the reactive, harmonic, and unbalanced load compensation is performed by auxiliary voltage source inverter (AVSII). This has an advantage that the rated capacity of MVSII can always be used to inject real power to the grid, if sufficient renewable power is available at the dc link.

In the DVSII scheme, as total load power is supplied by two inverters, power losses across the semiconductor switches of each inverter are reduced. This increases its reliability as compared to a single inverter with multifunctional capabilities. Also, smaller size modular inverters can operate at high switching frequencies with a reduced size of interfacing inductor, the filter cost gets reduced. Moreover, as the main inverter is supplying real power, the inverter has to track the fundamental positive sequence of current. This reduces the bandwidth requirement of the main inverter. The inverters in the proposed scheme use two separate dc links. Since the auxiliary inverter is supplying zero sequence of load current, a three-phase three-leg inverter topology with a single dc storage capacitor can be used for the main inverter. This in turn reduces the dc-link voltage requirement of the main inverter. Thus, the use of two separate inverters in the proposed DVSII scheme provides increased reliability, better utilization of microgrid power, reduced dc grid voltage rating, less bandwidth requirement of the main inverter, and reduced filter size. Control algorithms are developed by instantaneous symmetrical component theory (ISCT) to operate DVSII in grid-connected mode, while considering nonstiff grid voltage.

II. PROPOSED DUAL VOLTAGE SOURCE INVERTER

A. System Topology

The proposed DVSII topology is shown in Fig. It consists of a neutral point clamped (NPC) inverter to realize AVSII and a three-leg inverter for MVSII [18]. These are connected to grid at the PCC and supplying a nonlinear and unbalanced load. The function of the AVSII is to compensate the reactive, harmonics, and unbalance components in load currents. Here, load currents in three phases are represented by i_{la} , i_{lb} , and i_{lc} , respectively. Also, $i_g(abc)$, $i_{MVSII}(abc)$, and $i_{AVSII}(abc)$ show grid currents, MVSII currents, and AVSII currents in three phases, respectively. The dc link of the AVSII utilizes a split capacitor topology, with two capacitors C_1 and C_2 . The MVSII delivers the available power at distributed energy resource (DER) to grid. The DER can be a dc source or an ac source with rectifier coupled to dc link. Usually, renewable energy sources like fuel cell and PV generate power at variable low dc voltage, while the variable speed wind turbines generate power at variable ac voltage. Therefore, the power generated from these sources use a power conditioning stage before it is connected to the input of MVSII. In this study, DER is being represented as a dc source. An inductor filter is used to eliminate the high-frequency switching components generated due to the switching of power electronic switches in the inverters. The system considered in this study is assumed to have some amount of feeder resistance R_g and inductance L_g . Due to the presence of this feeder impedance, PCC voltage is affected with harmonics. Section III describes the extraction of fundamental positive sequence of PCC voltages and control strategy for the reference current generation of two inverters in DVSII scheme.

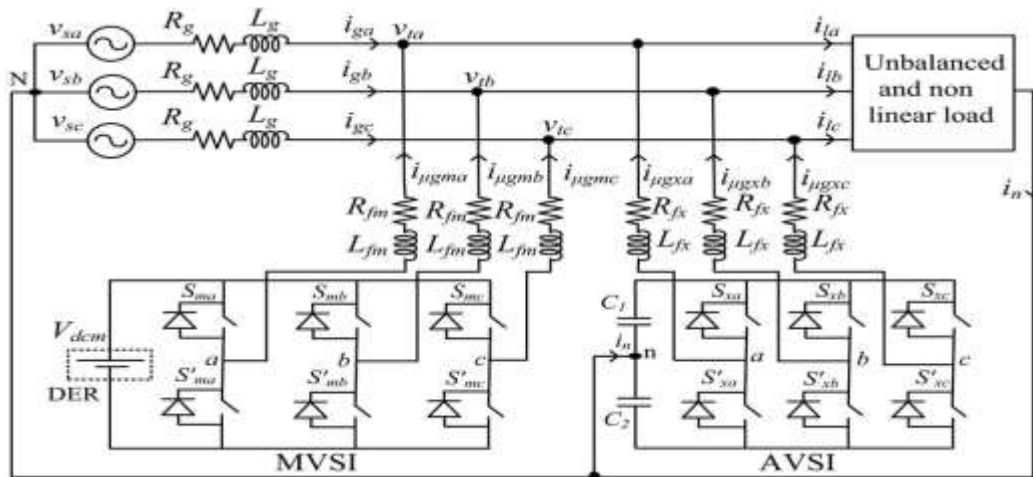


Fig.4 . Topology of proposed DVSI scheme.

Design of DVSI Parameters

1) AVSI:

The important parameters of AVSI like dc-link voltage (V_{dc}), dc storage capacitors ($C1$ and $C2$), interfacing inductance (L_{fx}), and hysteresis band ($\pm h_x$) are selected based on the design method of split capacitor DSTATCOM topology [16]. The dc-link voltage across each capacitor is taken as 1.6 times the peak of phase voltage. The total dc-link voltage reference ($V_{dc\text{ref}}$) is found to be 1040 V. Values of dc capacitors of AVSI are chosen based on the change in dc-link voltage during transients. Let total load rating is 5 kVA. In the worst case, the load power may vary from

minimum to maximum, i.e., from 0 to 5 kVA. AVSI needs to exchange real power during transient to maintain the load power demand. This transfer of real power during the transient will result in deviation of capacitor voltage from its reference value. Assume that the voltage controller takes n cycles, i.e., nT seconds to act, where T is the system time period. Hence, maximum energy exchange by AVSI during transient will be nST . This energy will be equal to change in the capacitor stored energy. Therefore

$$\frac{1}{2}C_1(V_{\text{dcr}}^2 - V_{\text{dc1}}^2) = nST$$

where V_{dcr} and V_{dc1} are the reference dc voltage and maximum permissible dc voltage across $C1$ during transient, respectively. Here, $S=5$ kVA, $V_{\text{dcr}} = 520$ V, $V_{\text{dc1}} = 0.8 * V_{\text{dcr}}$ or $1.2 * V_{\text{dcr}}$, $n = 1$, and $T = 0.02$ s. Substituting these values in (1), the dclink capacitance ($C1$) is calculated to be 2000 μF . Same value of capacitance is selected for $C2$. The interfacing inductance is given by

$$L_{fx} = \frac{1.6V_m}{4h_x f_{\text{max}}}$$

Assuming a maximum switching frequency (f_{max}) of 10 kHz and hysteresis band (h_x) as 5% of load current (0.5 A), the value of L_{fx} is calculated to be 26 mH.

2) MVSI:

The MVSI uses a three-leg inverter topology. Its dc-link voltage is obtained as $1.15 * V_{ml}$, where V_{ml} is the peak value of line voltage. This is calculated to be 648 V. Also, MVSI supplies a balanced sinusoidal current at unity power factor. So, zero sequence switching harmonics will be absent in the output current of MVSI. This reduces the filter requirement for MVSI as compared to AVSI. In this analysis, a filter inductance (L_{fm}) of 5 mH is used.

III. CONTROL STRATEGY FOR DVSI SCHEME

A. Fundamental Voltage Extraction

The control algorithm for reference current generation using ISCT requires balanced sinusoidal PCC voltages. Because of the presence of feeder impedance, PCC voltages are distorted. Therefore, the fundamental positive sequence components of the PCC voltages are extracted for the reference current generation. To convert the distorted PCC voltages to balanced sinusoidal voltages, $dq0$ transformation is used. The PCC voltages in natural reference frame (v_{ta} , v_{tb} , and v_{tc}) are first transformed into $dq0$ reference frame as given by

$$\begin{bmatrix} v_{td} \\ v_{tq} \\ v_{t0} \end{bmatrix} = C \begin{bmatrix} v_{ta} \\ v_{tb} \\ v_{tc} \end{bmatrix}$$

where

$$C = \sqrt{\frac{2}{3}} \begin{bmatrix} \sin \theta & \sin(\theta - \frac{2\pi}{3}) & \sin(\theta + \frac{2\pi}{3}) \\ \cos \theta & \cos(\theta - \frac{2\pi}{3}) & \cos(\theta + \frac{2\pi}{3}) \\ \frac{1}{\sqrt{2}} & \frac{1}{\sqrt{2}} & \frac{1}{\sqrt{2}} \end{bmatrix}$$

In order to get θ , a modified synchronous reference frame (SRF) phase locked loop (PLL) [23] is used. The schematic diagram of this PLL is shown in Fig. It mainly consists of a proportional integral (PI) controller and an integrator. In this PLL, the SRF terminal voltage in q -axis (v_{tq}) is compared with 0 V and the error voltage thus obtained is given to the PI controller. The frequency deviation $\Delta\omega$ is then added to the reference frequency ω_0 and finally given to the integrator to get θ . It can be proved that, when, $\theta = \omega_0 t$ and by using the Park's transformation matrix (C), q -axis voltage in $dq0$ frame becomes zero and hence the PLL will be locked to the reference frequency (ω_0).

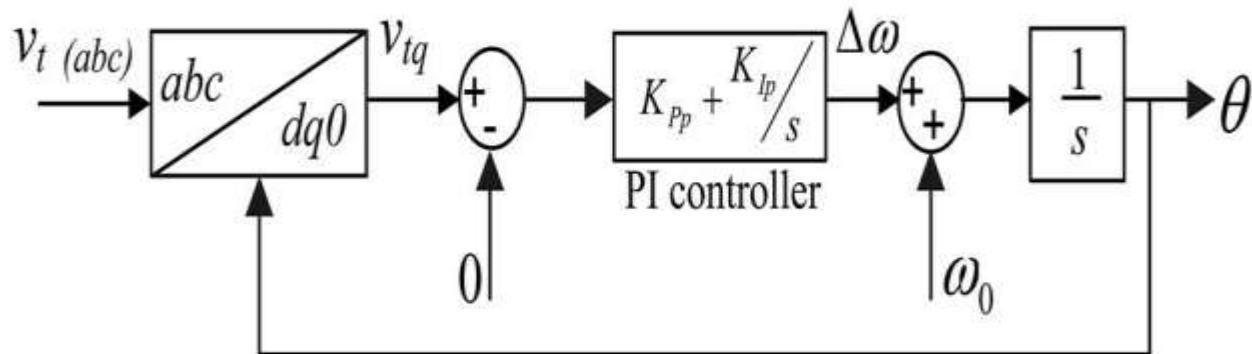


Fig.5 Schematic diagram of PLL.

B. Instantaneous Symmetrical Component Theory

ISCT was developed primarily for unbalanced and nonlinear load compensations by active power filters. The system topology shown in Fig

is used for realizing the reference current for the compensator. The ISCT for load compensation is derived based on the following three conditions

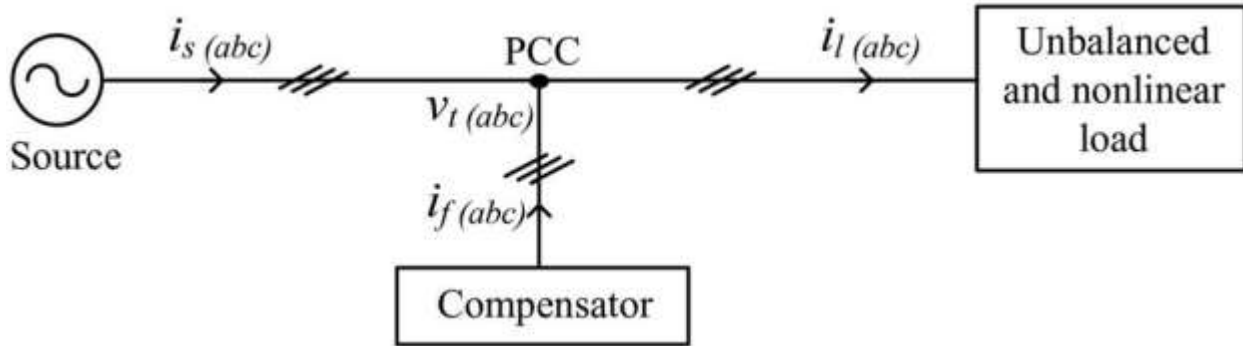


Fig.6. Schematic of an unbalanced and nonlinear load compensation scheme.

- 1) The source neutral current must be zero. Therefore

$$i_{sa} + i_{sb} + i_{sc} = 0. \quad (6)$$

- 2) The phase angle between the fundamental positive sequence voltage (v_{ta1}^+) and source current (i_{sa}) is ϕ

$$\angle v_{ta1}^+ = \angle i_{sa} + \phi. \quad (7)$$

- 3) The average real power of the load (P_l) should be supplied by the source

$$v_{ta1}^+ i_{sa} + v_{tb1}^+ i_{sb} + v_{tc1}^+ i_{sc} = P_l. \quad (8)$$

Solving the above three equations, the reference source currents can be obtained as

$$\begin{aligned} i_{sa}^* &= \left(\frac{v_{ta1}^+ + \beta(v_{tb1}^+ - v_{tc1}^+)}{\sum_{j=a,b,c} v_{tj}^{+2}} \right) P_l \\ i_{sb}^* &= \left(\frac{v_{tb1}^+ + \beta(v_{tc1}^+ - v_{ta1}^+)}{\sum_{j=a,b,c} v_{tj}^{+2}} \right) P_l \\ i_{sc}^* &= \left(\frac{v_{tc1}^+ + \beta(v_{ta1}^+ - v_{tb1}^+)}{\sum_{j=a,b,c} v_{tj}^{+2}} \right) P_l \end{aligned} \quad (9)$$

A modification in the control algorithm is required, when it is used for DVSI scheme. The following section discusses the formulation of control algorithm for DVSI scheme. The source currents, $i_s(abc)$ and filter currents $i_f(abc)$ will be equivalently represented as grid currents $i_g(abc)$ and AVSI currents $i_{\mu g \chi}(abc)$, respectively, in further sections.

IV. SIMULATION RESULTS

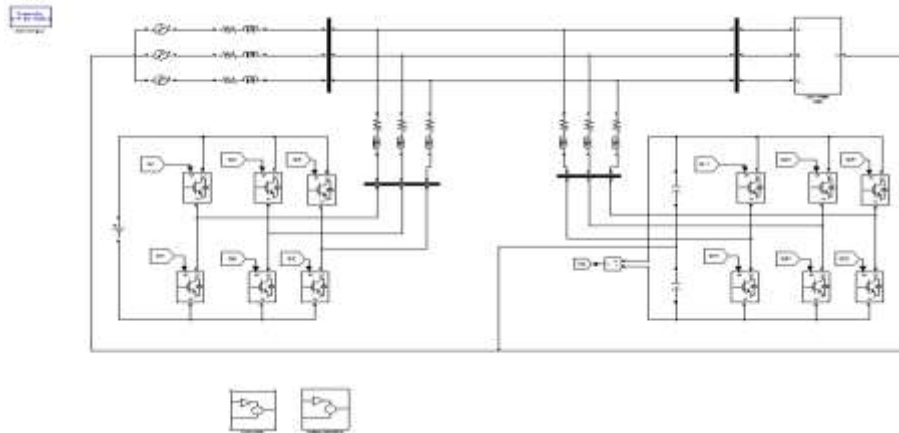


Fig 7 Matlab/simulink diagram of proposed DVSI system

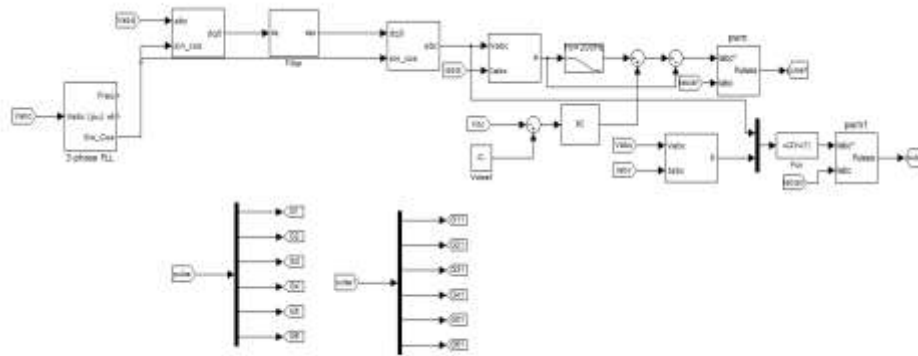


Fig 8 Controller SUBSYSTEM

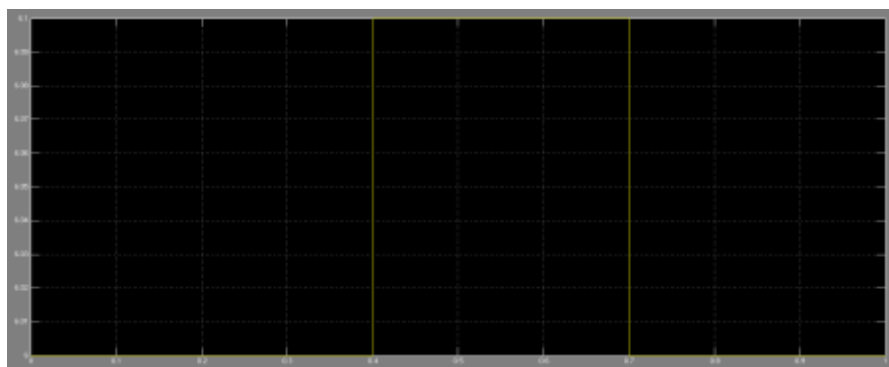


Fig 9 load active power

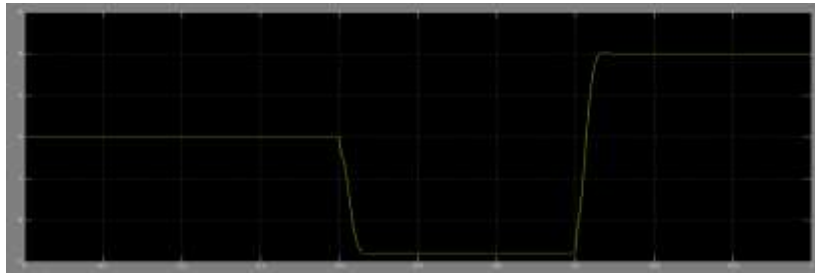


Fig 10 Active power supplied by grid

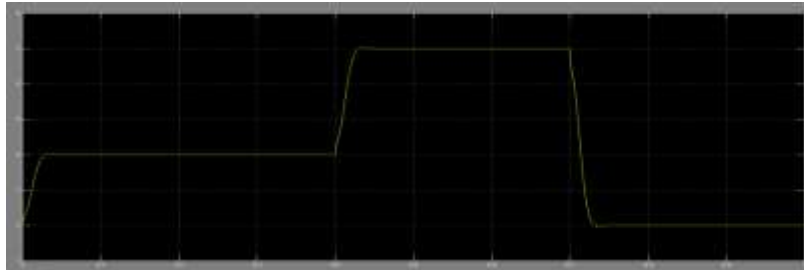


Fig 11 active power supplied by MVSI;

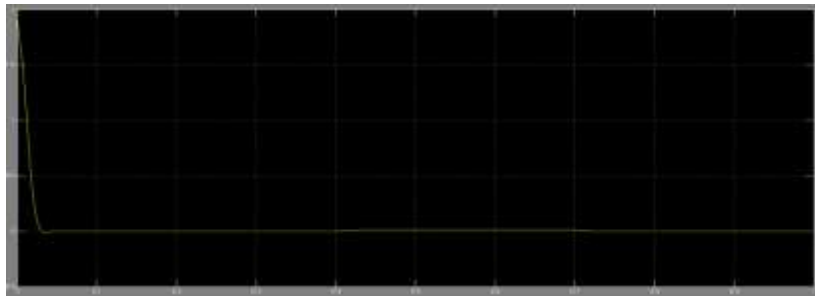


Fig 12 active power supplied by AVSI.

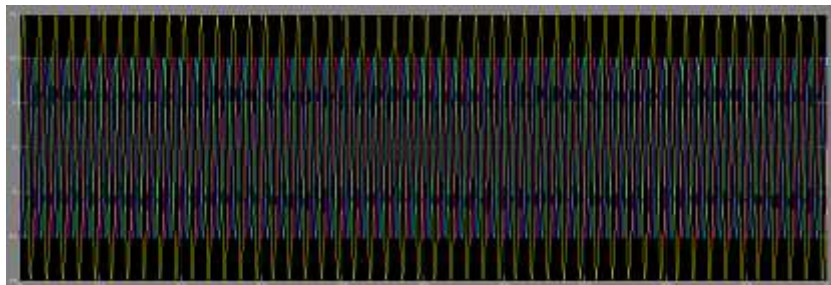


Fig 13 load currents

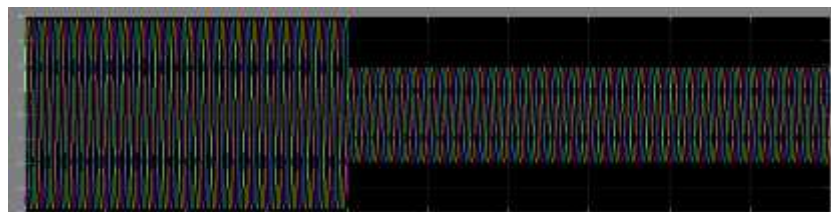


Fig 14 grid currents

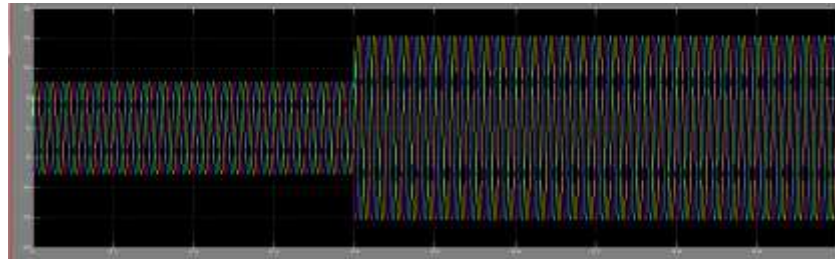


Fig 15 MVSII currents

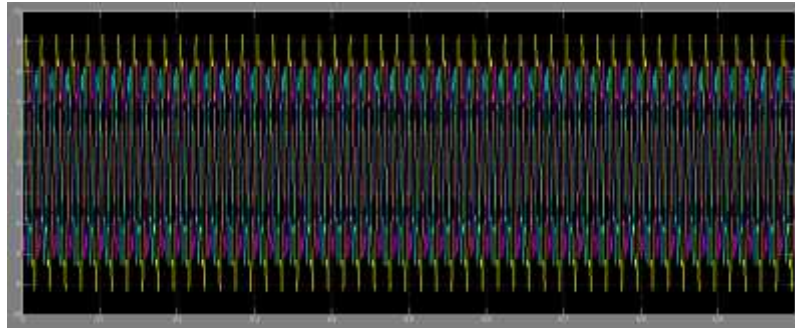


Fig 16 AVSI currents

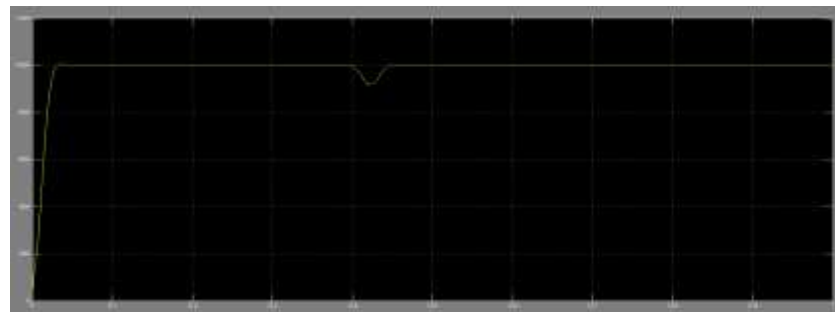


Fig 17 DC Link voltage

V.CONCLUSION

The proposed DVSI Control algorithms are developed to generate reference currents for DVSI using ISCT. The proposed scheme has the capability to exchange power from distributed generators (DGs) and also to compensate the local unbalanced and nonlinear load. The performance of the proposed scheme has been validated through simulation and experimental studies. As compared to a single inverter with multifunctional capabilities, a DVSI has many advantages such as, increased reliability, lower cost due to the reduction in filter size, and more utilization of inverter capacity to inject real power from DGs to microgrid. Moreover, the use of three-phase, threewire topology for the main inverter reduces the dc-link voltage requirement. Thus, a DVSI scheme is a suitable interfacing option for microgrid supplying sensitive

loads. Simulation results are observed for a proposed fuzzy based DVSI.

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