

# A Study and Design of Circulating Current and DC Current Ripple Controlling MMC

**S.Vijay Kumar**

M.Tech(Digital Systems And Computer Electronics), JNTUH

**R.Srikanth**

M.E.(Bio Medical Engineering), UCEOU

**Vuppari Prabhakar**

M.Tech(Energy Systems), JNTUCEA

**ABSTRACT:** MMC is an evolving topology which is able to handle high voltage and power ratings. Under the unbalanced condition, the main theme of control is to decrease the negative-sequence line current. Moreover, the circulating current of a MMC contains not only double-line negative-sequence component, which appears under the balanced condition, but also positive- and zero-sequence double-line components. Thus the controller should be able to remove all those components of the circulating current. A proportional-resonant controller (PR controller) is applied to regulate the positive- and negative-sequence components of circulating current. In addition, the zero-sequence component of circulating current is controlled by a dc current controller. The proposed control method is confirmed in MATLAB/SimPowerSystem.

**KEYWORDS:** Modular multilevel converter (MMC), Circulating Current Suppression Control, DC Current Ripple Control, Proportion Resonance Controller

## I. INTRODUCTION

AC has been the desired international platform for electrical transmission to homes and corporations for the beyond 100 years. And but high-voltage AC transmission has some limitations, beginning with transmission potential and distance constraints, and the impossibility of immediately connecting two AC electricity networks of various frequencies. With the dawn of a brand new energy era and the want to construct a smarter grid, HVDC is predicted to develop some distance past its

traditional function as a complement to AC transmission [2].

High Voltage Direct Current (HVDC) is the electric electricity transmission choice applied in large quantity of power over lengthy distances with minimal losses. Considering the reality that during a traditional three phase gadget the strength brought is conformed by its RMS price, HVDC allows transmitting active electricity with higher voltage variety. Moreover, the impedance created in AC transmission structures are avoidable lowering the energy losses. Therefore, the preliminary set up price of HVDC is better than HVAC structures but because of decrease losses it turns into value powerful over the time. For instance, strength brought from faraway offshore wind farms may be correctly fed into strength grids onshore via HVDC generation. Moreover, HVDC structures are useful to interconnect asynchronous AC grids reliably. Using HVDC system lets in the opportunity of using underground and subsea cables. Hence, HVDC is taken into consideration as a distinctly green alternative for transmitting huge amounts of strength over lengthy distances and for special cause applications. As a key enabler within the destiny strength system based on renewables, HVDC is surely shaping the grid of the future. This era consists of a converter station in which the AC system is converted into DC then transmitted via a strength transmission cable and then converted back into AC. The cable connection may be overhead or each underground or submarine beneath water. An HVDC transmission system is depicted in Figure 1.1.

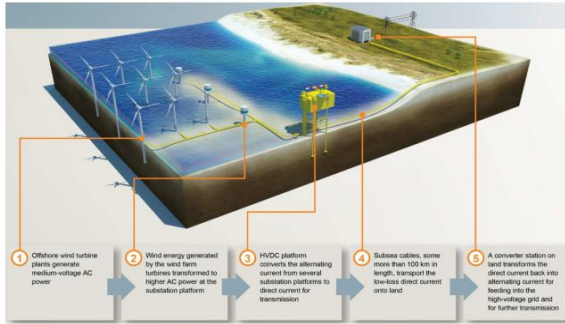


Figure 1.1: HVDC transmission system from an offshore to onshore grid [1]

Modular Multilevel Converters (MMCs) have won researcher's attention due to their potential to deal with excessive voltage and energy ratings. VSC-HVDC is getting increasingly more important for integrating renewable power assets together with large offshore wind farms, offering bendy interconnection among vulnerable AC grid network the use of back-to-back configuration, or truly transmitting electricity using underground cables. The VSC-HVDC additionally has speedy and particular manage over the lively strength-flow in addition to it could independently control the reactive power injection on the nearby AC grid. There are numerous operational MMC-HVDC tasks which include HVDC PLUS (Siemens) with an 88km undersea transmission hyperlink between San Francisco's City Centre electric power grid and a substation near Pittsburg. The important assisting functions HVDC PLUS affords are AC voltage Control, black-start functionality, compact converter station space utilization, four quadrant operation, repayment of asymmetrical masses, bendy integration into HVDC multi terminal structures or future HVDC grids. Its primary running principle and different blessings each on the technical in addition to on the budget friendly thing may be described in [3] [4].

Another MMC-HVDC set up named HVDC Light by using ABB, is a version of HVDC classic used to transmit strength in strength tiers (50-2500MW) transmitted using overhead lines and environmental pleasant underground and sub-sea cables. It is used for grid interconnections and offshore links to wind farms. With HVDC Light, it's far feasible to transmit energy in both instructions and to guide existing AC grids a good way to increase robustness, stability, reliability and

controllability. HVDC Light gives many other advantages and may be utilized in extraordinary programs which is explained in [5]. As mentioned before, the principle trouble of the 2 degree converter is its excessive switching losses due to exceedingly high switching frequency which necessitates excessive insulation necessities of the transformer, as well as filters. The use of modular multilevel converters overcomes most of the aforementioned shortcomings, but on the expense of two times as many semi-conducting devices and a big allotted capacitor for each submodule. The precept idea of the hybrid VSC-HVDC, as used in HVDC MaxSine evolved by Alstom, is to use a stage converter as the principle switching aspect with low switching frequency and an MMC to provide a voltage wave shaping function on the AC side so that you can do away with the harmonics [6] [7].

## II. PROPOSED CONTROL STRATEGY

Multilevel converters are classified into diode-clamped multilevel converters (DCMCs), flying-capacitor multilevel converters (FCMCs), cascaded H-bridge converters (CHBCs), and modular multilevel converters (MMCs). MMCs have been widely adopted in VSC-HVDC systems. Fig. 1 shows the structure of an MMC consisting of six arms. Each arm is composed of an inductor and a series of connected half-bridge submodules (SMs). HVDC-MMC systems require several design techniques. (1) System parameter design includes inductance and capacitance design and switching device current capacity design. (2) System control design includes power (DC-link voltage) control, AC-side current control, circulating current control, and SM voltage balancing [2-6].

A design method for the SM capacitance of the MMC was introduced in [7]. This design method calculated the difference in input energy according to the amplitude of the grid voltage and the active power. The SM capacitance is designed by the input energy and the SM capacitor voltage ripple on the basis of the limit value.

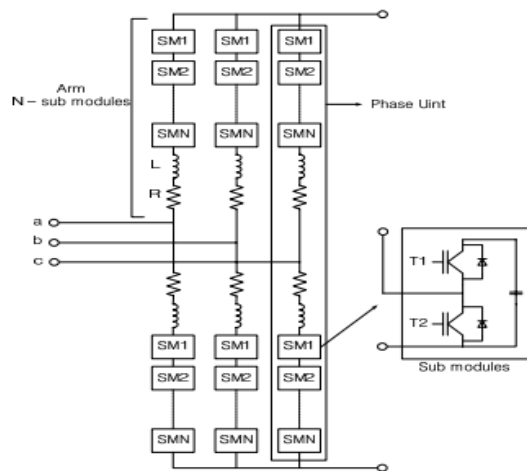


Fig. 1. Basic structure of MMC.

The SM capacitor voltage ripple has line-frequency and double-line-frequency components. However, this design method did not separate line-frequency and double-line-frequency components; the capacitor voltage ripple was only calculated using integrated components.

The output of sub-module is either  $V_c$  or 0 depending on the gate statement. When  $N$  is big enough or the switching frequency is high enough, the voltage injection to each arm by sub-modules can be considered as continuous. For DC side voltage, with big DC-side capacitor, the dc-side voltage can be considered as a constant value. Thus, the single phase equivalent circuit of a MMC can be expressed as Fig. 2.

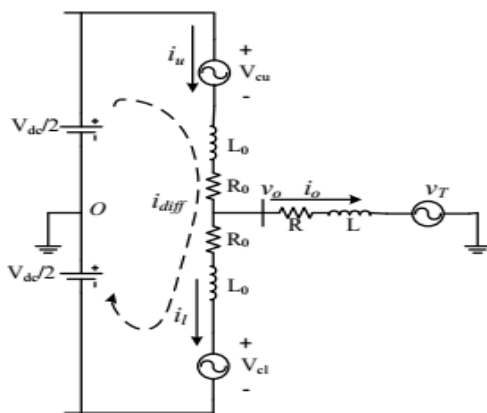


Fig. 2. Single phase equivalent circuit of MMC

In Fig. 2, The upper and lower arm current are named as  $i_u$  and  $i_l$ ; the converter's output current and voltage are

named as  $i_o$  and  $v_o$  respectively. The circulating current flowing within the converter is denoted as  $i_{diff}$ . Since the upper and lower arm are symmetric, ideally both lower and upper arm currents contain half of the converter output current.

### III. CONTROL SYSTEM UNDER UNBALANCED VOLTAGE

#### A. PR Controller

PR control can achieve high bandwidth at certain resonant frequency. Through PR control, measurement signal can track the reference signal without steady-state error at the resonant frequency.

#### B. Outer Loop Power Control

When the grid side voltage is under unbalanced condition, the line current and power flow are separated in positive-, negative- and zero-sequence components. With a zero-sequence current controller, the zero-sequence current can be reduced to zero. Therefore the objective of the unbalance controller is the negative component of line current.

#### C. Inner Loop Current Control

Different control objectives have been set for MMC under unbalanced condition. Reference [15] tried to reduce the negative components of the line current to zero.

#### D. Grid-side Zero-Sequence Current Control

The overall control structure is shown in Fig. 4. The zero sequence current occurs during the unbalanced condition. A Y-to- $\Delta$  transformer can stop the zero sequence current. However, when the fault happens on the transformer or between the transformer and MMC, zero-sequence current will not be stopped by transformer.

#### E. Circulating Current Control and Dc Current Ripple Control

In an MMC, the difference between each phase's total sub-module capacitor voltage leads to the circulating current. Under balanced condition, it has been evaluated that circulating current consists of only negative-sequence double line frequency component, since the instantaneous

power for each phase has a negative-sequence double line frequency component.

When the ac-side voltage has negative component, the instantaneous power of each phase consists not only negative sequence but also positive- and zero-sequence double-line frequency components. Therefore, to eliminate the circulating current under unbalanced condition, the controller in positive-, negative- and zero-sequence are all needed.

The dc component is set as the reference of the PR controller. The output of PR controller is the positive- and negative-sequence components of the reference inner unbalanced voltage, which is noted as  $e_{diff,abc}$  in Fig. 3.

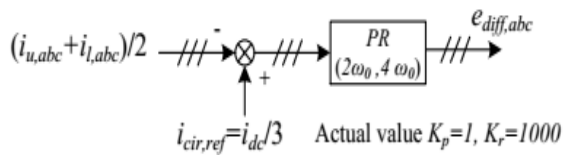


Fig. 3. Circulating current suppression controller.

For a three-phase system, the sum of positive- and negative sequence current are zero. However the sum of zero-sequence component is not zero. And  $i_{dc}$  is the sum of three-phase current. Therefore, if the three-phase current has zero-sequence component, then  $i_{dc}$  includes the zero-sequence component ripple. Normally, a Y-to- $\delta$  transformer can stop the zero sequence line current. However, when the fault happens between the transformer and MMC, or the system has no transformer, it is necessary to eliminate the zero-sequence line current. The controller to eliminate zero-sequence current is shown in Fig. 4. As shown in Fig. 3, the output of the zero sequence line current controller is added to the output of the inner loop current controller as a zero-sequence component.

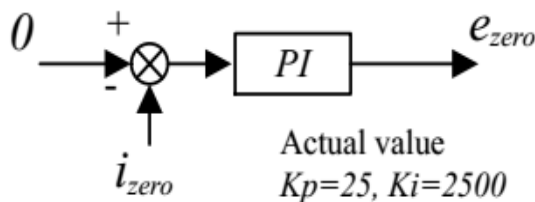


Fig. 4. Zero-sequence line current controller.

If we assume there is no power loss on MMC. Then the ac-side power of MMC should be equal to the dc-side power of MMC. So the dc component of the dc-side current can be easily set as  $I_{dc,ref} = P_{out}/V_{dc}$ . Where  $P_{out}$  is the ac-side output power of MMC and  $V_{dc}$  is the dc supply voltage of MMC. A PR controller is used to control the zero-sequence of the inner difference current to  $I_{dc,ref}$ . The output of PR controller is the zero-sequence component of the reference inner difference voltage. Beside the double-line frequency ripple, the dc current ripple controller can also cancel the resonance current caused by LC circuit resonant. A controller is added to reduce the dc current ripple as shown in Fig. 5. The output of the dc current ripple controller is added to the output of the circulating current suppression controller as shown in Fig. 3.

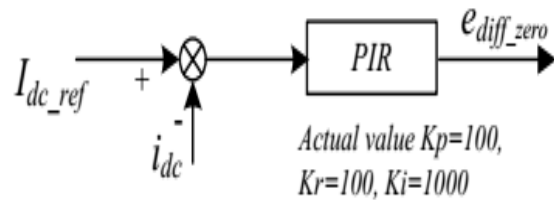


Fig. 5. DC-side current ripple controller.

#### IV. SIMULATION RESULTS

A simulation of the proposed system is conducted in MATLAB SimPowerSystem. The simulation environment and parameters are listed in table I. At  $t=0:2s$  the circulating current controller and the controller to eliminate the dc current ripple is activated. From  $t=0:6s$  to  $t=0:8s$ , there is a 0.2pu negative-sequence component voltage on the grid side. Regardless of the start-up process, the capacitor voltages of sub modules were charged at nominal value at beginning.

#### B. Performance of the control system

Fig. 6 shows the output real power of the MMC. During the unbalanced voltage condition ( $t$  from 0:6 to 0:8), the output power has a double-line frequency component. When  $I^-$  is zero, with a non zero  $V^-_g$ , The double-line frequency component of line power is not zero.

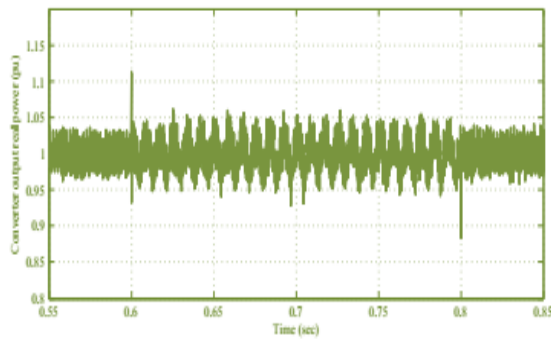


Fig. 6. Output real power of converter.

Fig.7 shows the output reactive power of the MMC. As the same with the output active power. During the unbalanced grid condition, there is a double-line frequency component appear on the converter output reactive power.

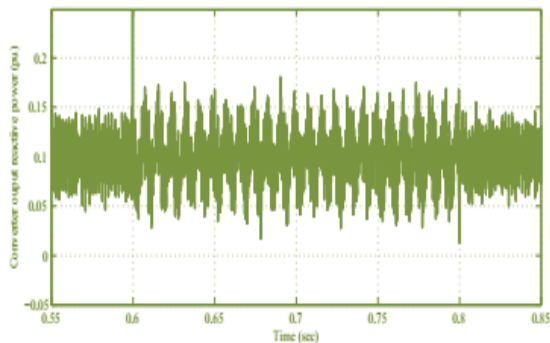


Fig. 7. Output reactive power of converter

Fig.8 is the grid current of MMC during the unbalance condition. The ripple during the unbalanced condition is due to the changing of the power and reactive power. The outloop power controller tries to regulate the converter power to reference value. So the ripple of the power leads a ripple in current reference in dq frame, and the ripple for reference current in dq frame causes a ripple of line current in abc frame.

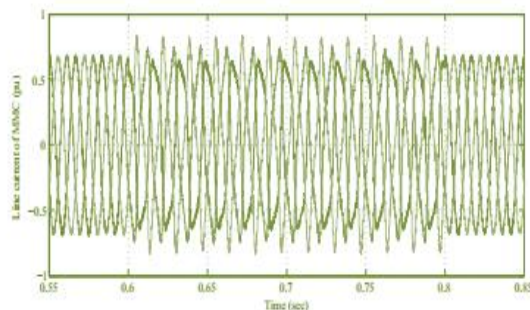


Fig. 8. Grid current of MMC

Fig. 9 is the zero-sequence component of line current. After applying the zero-sequence current controller at  $t = 0.1s$ , the magnitude of the zero-sequence current is reduced during both balanced and unbalanced condition.

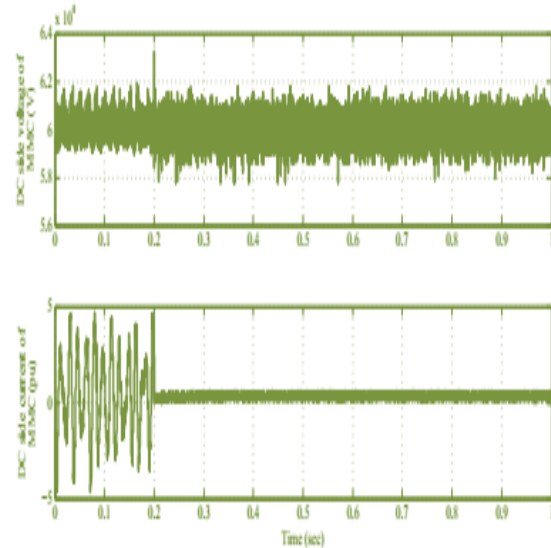


Fig. 9. Zero-sequence component of line current of MMC

## V. CONCLUSION

Investigative study of the internal unbalances in an MMC has been approved out. This unbalance can be a result of the asymmetries, non linearities and change in the tolerances of the components as for illustration arm inductors and submodule capacitors used in an MMC. When the negative-sequence component of the line current during unbalanced condition is controlled to zero, there is a double-line frequency component on the output power of MMC during unbalanced condition. With proper nonzero negative-sequence line current, the double-line frequency component of the output real power can be eliminated. The PR controller reduces both positive- and negative-sequence components of the circulating current during the unbalanced condition. In addition, a dc current controller is applied to reduce the zero-sequence of the circulating current and resonance current on the dc-side of MMC.

## REFERENCES

- [1] Siemens, "HVDC converter station," <http://www.pennenergy.com/articles/pennenergy/2014/04/siemens-wins-borwin3-north-sea-power-transmission-order.html>

[2] ABB, “HVDC over HVAC,” <http://new.abb.com/systems/hvdc/why-hvdc>.

[3] Siemens, “HVDC PLUS VSC Technology,” <http://www.energy.siemens.com/mx/en/power-transmission/hvdc/hvdc-plus.htm#content=%20Topology>.

[4] “Basics and principle operation of HVDCPLUS,” [http://www.energy.siemens.com/br/pool/hq/power-transmission/HVDC/HVDC\\_Plus\\_Basic%20and%20Principals.pdf](http://www.energy.siemens.com/br/pool/hq/power-transmission/HVDC/HVDC_Plus_Basic%20and%20Principals.pdf).

[5] ABB, “HVDC Light, It’s time to connect,” <http://new.abb.com/docs/default-source/ewea-doc/hvdc-light.pdf?sfvrsn=2>.

[6] Alstom, “HVDC-VSC transmission technology of the future,” <https://www.gegridsolutions.com/alstomenergy/grid/Global/Grid/Resources/Documents/Smart%20Grid/Think-Grid-08-%20EN.pdf>.

[7] M. Winkelkemper, A. Korn, and P. Steimer, “A modular direct converter for transformerless rail interties,” in Industrial Electronics (ISIE), 2010 IEEE International Symposium on, July 2010, pp. 562–567.

[8] Q. Tu, Z. Xu, and L. Xu, “Reduced switching-frequency modulation and circulating current suppression for modular multilevel converters,” Power Delivery, IEEE Transactions on, vol. 26, no. 3, pp. 2009–2017, July 2011.

[9] L. Harnefors, A. Antonopoulos, S. Norrga, L. Angquist, and H.-P. Nee, “Dynamic analysis of modular multilevel converters,” Industrial Electronics, IEEE Transactions on, vol. 60, no. 7, pp. 2526–2537, July 2013.

[10] Y. Ma, L. Fan, and Z. Miao, “Integrated control and switching strategy for a grid-connected modular multilevel converter,” in PES General Meeting, 2015 IEEE, accepted, July 2015, pp. 1–5.

[11] X. She, A. Huang, X. Ni, and R. Burgos, “AC circulating current suppression in modular multilevel converter,” in IECON 2012 - 38<sup>th</sup> Annual Conference on IEEE Industrial Electronics Society, Oct 2012, pp. 191–196.

[12] Y. Ma, Z. Miao, V. R. Disfani, and L. Fan, “A one-step model predictive control for modular multilevel

converters,” in PES General Meeting Conference Exposition, 2014 IEEE, July 2014, pp. 1–5

[13] L. Fan, R. Kavasseri, H. Yin, C. Zhu, and M. Hu, “Control of DFIG for rotor current harmonics elimination,” in Power & Energy Society General Meeting, 2009. PES’09. IEEE. IEEE, 2009, pp. 1–7.

[14] L. Fan, H. Yin, and Z. Miao, “A novel control scheme for DFIG-based wind energy systems under unbalanced grid conditions,” Electric Power Systems Research, vol. 81, no. 2, pp. 254–262, 2011.

[15] M. Guan and Z. Xu, “Modeling and control of a modular multilevel converter-based hvdc system under unbalanced grid conditions,” Power Electronics, IEEE Transactions on, vol. 27, no. 12, pp. 4858–4867, Dec 2012.

#### Authors Details:



S. Vijay Kumar completed M.Tech (Digital Systems And Computer Electronics) from JNTUH



R. Srikanth completed M.E. (Bio Medical Engineering) from UCEOU.



Vuppari Prabhakar Completed M.Tech (Energy Systems) from JNTUCEA.