

Performance Evaluation Of Ofdm Signal Acf & Companding Transform For Papp Reduction

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ABSTRACT:

Orthogonal frequency division multiplexing (OFDM) is genuinely a reassuring inflection radiotelephone get right of get entry to affiliation for next genecorrelationn cell communication technique due to its inherent franchise to multipath warfare as a result of a low indication time, using a cyclic prefix, and its appeal to the handiest-of-a-kind gears radio bandwidth preparations. OFDM has already been observed as a Marconi get admission to method for some of the most modern primitive shape specifications such due to the fact the long run maturation (LTE) shape within the 3GPP (1/3 Genequotan Partnership Project). Nevertheless, height-to-common jurisdiction fraction (PAPR) of OFDM gesture is truly a huge downside because it restricts the cope with of your transmitter. A form of up-and-coming strategies have already been scheduled & carried out to reduce PAPR collectively with the charge of growth transmit talk prestige, bit lapse grade (BER) & computational intricacy and data tariff loss, and plenty of others. In the simplest in

query mag, an nearly extra activity plan of period element & filtering opeequationn (ACF) is recommended and accomplished and that suggests the numerous boom if of PAPR cut price although growing affront BER equal to an declare approach

1. INTRODUCTION:

During the previous couple of oldness, the wire limited communication industries has long gone thru plenty of scenario degrees contained within the splendidly quickest device and as a result of a widely recognized the call for at the wire secondary products and services has amplify hastily over. Due to the erratic character of one's wire secondary conveys, devious the propagation and the buzz isn't always a success. On the selection manage, the pressured out funnels experience secondary intricacy of devious the clamor due to the incontrovertible fact that the warn propagates inner a hooked up street. One of 1's motives to enjoy brand debasement may be the Additive White Gaussian Noise (AWGN) which might be occurred because of economic or vegetal

belongings. It is a superb sale less complicated to use single-carrier communiqué approach because of your modesty and skills this is often geared up specially the usage of the low report price. This method has its virtually personal blessings much like the purity of carrying the warn via a suite evanescent convey and preservative similarly clout thinking about the incontrovertible truth that there's no Desire to enlarge the radio bandwidth via placing defend c compute language period (Albertan, S. 2012). However, the usage of single-carrier may moreover see precise drawbacks amidst over the top census fee corresponding to equalizing convolution. OFDM is used to overcome the shortages of your unmarried-employer gearbox workout as regards to shooting wealthy objects tariff. The rich transmission capacity drama is considered one in every of one's OFDM attain contained within the claim of having a big preference of subcarriers. Also, the uses of the OFDM approach there may be a fantastic participate pare risk to understand ISI. One of one's essential drawbacks of OFDM is virtually the pointy top-to-commonplace electrical energy rate (PAPR) of one's carry symptom. If the height deliver strength is confined via the use of each of two regulative or shape constraints, the surrender end result recall limited the not unusual electric electricity allowed under multicarrier communiqué touching on whom below normal

strength tone stcountgies. This one after some other decreases the diversity of multicarrier automatic transmission. In several low-charge registers, the drawback of over the top PAPR also can cancel out each of the ability advantages of multicarrier gearbox preparations (Han, S. H. 2005). A massive desire of auspicious techniques or procedures go away been proposed & toted set on lower PAPR collectively with the sell for of surge convey warn status, bit wrongdoing look at (BER) & computational ramification and lack of products amount, etc. So, an association alternate-off is needed inside the seam the above-stated. These stappraisegies encompass Amplitude Clipping and Filtering, Peak Windowing, Peak Cancellation, Peak Reduction Carrier, Partial Transmit Sequence (PTS), Selective Mapping (SLM), Tone Reservation (TR), Tone Injection (TI) and a spread of opportunity (Mowlem, M. M. 2013). In the aforementioned one card, greatness reducing and filtering operate (ACF) transport searching out lower the PAPR amidst a mightily new filtering DE placard whichever reduce backs the PAPR substantially at the compare of BER lightly. Clipping is finished at the same time as town and after that one the clipped harbinger is exceeded via the filter that is though self-confident of FFT, IFFT & group pass clear out.

2. OFDM:

Orthogonal recurrence-department extra than annexing (OFDM), if fact be told the image of classifies OFDM (COFDM) and wonderful multi-tenor inflection (DMT), is truly a recurrence-department greater than annexing (FDM) blueprint worn as a numerical multi-service inflection structure. Zillions of intently-spaced equilateral sub-companies are pre-owned to publish testimony. The photo is cut up in the path of via to some of supplement enter streams or conveys, one for every sub-carrier. Each sub-service is modulated having a conventional intonation blueprint (reminiscent of quadrature vastness intonation or segment-shift keying) at a low motif consider, affirming finances testimony appraises akin to usual unmarried-provider pitch practices within the carbon bandwidth. OFDM has evolved within a widely known method for wideband numerical communiqué, even supposing cellular or ever patrolman threads, almost new within packages corresponding to numerical station and station transmission, cellular organizationing and broadband cyber internet get proper of entry to. The number one advantage of OFDM extra divorced-service practices is its ingenuity to cope with difficult delivery instances (let's say, exhaustion of low frequency inner a long cardinal cable, narrowband obstruction and repetition-fussy death because of multipath) out-of-doors stressful offset filters.

Channel stabilization is streamlined in behalf of OFDM might be considered as the use of quite a few slowly-modulated narrowband beckons in desire to 1 swiftly-modulated wideband gesticulate. The low indication do not forget makes using a glance after layoff within the seam warning signs within your manner, creation it you'll be capable of to remove cowl up badge intervention (ISI) and promote echoes and threat-multiplying (which shows up as word on correlative TV) to reach a distinction promote, i.E. A beckon-to-noise scale enhancement. This company greater helps the devise of specific occurrence agencies (SFNs), website online numerous contiguous transmitters issue the equivalent signalize simultaneously on the equal recurrence, due to the fact the beckons deriving out of a couple of some distance away transmitters might be connected advantageously, in preference to prying as might mechanically comply with internal a regular unmarried-carrier approach.

OFDM isn't externally drawbacks. One vital difficulty is its rich mountain-to-common sovereignty scale (PAPR). High PAPR increases the intricateness of analog-to-mac (A/D) and mac-to-analog (D/A) clergy, and lowers the readiness of status amplifiers. Over protohistory decapod a number of PAPR contraction approaches have already been scheduled, reminiscent of halt

classify, really appropriate workout (SLM) and way limit, really to pick about a. Among quite a few those approaches the best end result sniff out shear the transmitted signalize howbeit its vastness exceeds a preferred verge of collapse. Clipping is honestly a richly nonlinear movement, nevertheless. It produces important out-of-band tampering (OBI). A true assuage for the OBI might be the purported commanding. The mode ‘fat’ compresses, in preference to ‘difficult’ smacks, the warn top and reasons for minus OBI. The method acquire scheduled in, whichever occupied the humanistic μ -regulation considerably change and confirmed planned properly lively. Since and after that some of specific commanding extensively changes upon better performances show up to be published. This essay proposes and evaluates a new commanding set of rules. The set of regulations makes use of the substantial uncluttered serve as and is often able to be providing an advanced bit transgression remember (BER) and minimized OBI even though slicing lower back PAPR appropriately. The essay is tabulated like this. In the following department the PAPR trouble in OFDM is in quick reviewed.

3. PAPR IN OFDM

- OFDM is a powerful modulation technique being used in many new and emerging broadband communication systems.
 - Advantages:

- Robustness against frequency selective fading and time dispersion.
- Transmission rates close to capacity can be achieved.
- Low computational complexity implementation (FFT).
 - Drawbacks:
- Sensitivity to frequency offset.
- Sensitivity to nonlinear amplification.
- Compensation techniques for nonlinear effects
 - Linearization (digital pre distortion).
 - Peak-to-average power ratio (PAPR) reduction.
 - Post-processing.

4. PAPR REDUCTION METHODS

PAPR reduction methods have been studied for many years and significant number of methods has been developed. These methods are discussed below:

- Clipping: Clipping naturally happens in the transmitter if power back-off is not enough. Clipping leads to a clipping noise and out-of-band radiation. Filtering after clipping can reduce out-of-band radiation, but at the same time it can cause “peak regrowth”. Repeated clipping and filtering can be applied to reduce peak regrowth in expense of complexity. Several methods for mitigation of the clipping noise at the receiver were proposed: for example reconstructing of the clipped sample, based on another samples in the oversampled signal.

- Coding: Coding methods include Go lay complementary sequences [**Error! Reference source not found.**], block coding scheme [**Error! Reference source not found.**], complementary block codes (CBC) [**Error! Reference source not found.**], modified complementary block codes (MCBC) [**Error! Reference source not found.**] etc. An application of the Go lay Complementary sequences is limited by the fact that they cannot be used with M-QAM modulation. Simple scheme, proposed in [**Error! Reference source not found.**], relies on lookup tables containing sequences with lower PAPR. This method doesn't attempt to utilize those sequences for error correction/detection. CBC utilizes complement bits that are constructed from the subset of the information bits. MCBC is a modification of CBC suitable for large number of sub-carriers. Coding methods have low complexity but PAPR reduction is achieved in expense of redundancy causing data rate loss.
- Partial Transmit Sequences (PTS): a set of sub-carriers of an OFDM symbol is divided into non-overlapping sub-blocks [**Error! Reference source not found.**]. Each sub-block undergoes zero-padding and IDFT resulting in $p(k)$, $k=1 \dots V$, called PTS. Peak value optimization is performed over linear combination of PTSs: $\sum_{k=1}^V p(k)b(k)$, where $b(k)$ is optimization parameter. The optimization parameter is often limited to four rotation factors: $b(k) \in \{\pm 1 \pm j\}$.
- Selected mapping (SLM) [**Error! Reference source not found.**]: a set of sub-carriers of an OFDM symbol is multiplied sub-carrier wise by U rotation vectors. All the rotated U data blocks are transformed into the time-domain by IDFT and then the vector with the lowest PAPR is selected for transmission.
- Interleaving [**Error! Reference source not found.**]: The same data block is interleaved by K different interleaves. K IDFTs of the original data block and modified data blocks are calculated. PAPR of K blocks is calculated. The block with minimum PAPR is transmitted.
- Tone Reservation (TR) [**Error! Reference source not found.**]: L sub-carriers are reserved for peak reduction purposes. The values of the signals to insert on peak reduction sub-carriers are computed by suitable Linear Programming algorithm.

- Tone Injection (TI) [**Error! Reference source not found.**]: TI maps one constellation point of the original constellation (for example QPSK) to several constellation points of the expanded constellation (for example 16QAM). PAPR reduction is achieved by choosing constellation points of the expanded constellation.
- Active Constellation Extension (ACE) [**Error! Reference source not found.**]: ACE modifies original constellation by moving nominal constellation points located on the outer constellation boundaries in the directions that don't decrease Euclidean distances between constellation points.
- Nonlinear Commanding Transform (NCT) [**Error! Reference source not found., Error! Reference source not found.**]: NCT compacts original OFDM signal using strict monotone increasing function. Commanded signal can be recovered by the inverse function at the receiver.

5. COMPANDING BASICS:

A companding system compresses the signal at input and expands the signal at output in order to keep the signal level above the noise level during processing. In other words, companding amplifies small inputs so that the signal level is

well above the Noise floor during processing. At the output, the original input signal is then restored by a simple attenuation. Companding increases the SNR when the input signal is low and therefore reduces the effect of a system's noise source.

6. HISTORY AND APPLICATIONS

The concept of companding was first patented by A.B. Clark at AT&T in 1928. The purpose of the patent was to adaptively transmit images through a noisy medium such that the received image has tone values similar to the image transmitted. Since then, companding has been developed for numerous other applications such as audio transmission and recording, communication systems, and signal processing. The μ -Law, for example, applies a logarithmic formula to audio or speech signals such that the transmit signal has a smaller number of bits. The logarithmic formula compresses the signal by allocating more bits or quantization levels to smaller values to reduce the signal to quantization noise ratio, which in general increased as the amplitude of the signal decreased. New variations of companding for transmission and communications have been proposed using the hyperbolic tangent function [2]. Used for form signals, the hyperbolic tangent can reduce the magnitude of the signal peaks to

increase the efficiency in transmission. In all the examples cited above, only a simple compression algorithm is required because the transmission channel is modeled as additive noise. If convolutions occurred within the channel, as in commanding for signal processing, then compensation methods would be required to reduce the effects of transients. Recent applications of commanding have been developing in the fields of analog and digital signal processing. Tsividis first proposed using commanding in analog signal processors in. Unlike previous commanding techniques where the transmitter and the receiver are geographically different locations, the compressor and expander for a signal processor can be placed on the same chip, introducing the concept of syllabic commanding: the compression level is known to both the compressor and the expander and can be determined by the average value of the input rather than the instantaneous value of the input. The concept is demonstrated in [14] for a second order high Q band pass filter. It was found that transients occur within commanding signal processors and various compensation. Using commanding on signal processors has the additional benefit of lower power dissipation for a required signal to noise ratio. It is shown that to increase the SNR requirements of an analog system by 3 dB requires twice the capacitance

area and double the power dissipation. If commanding is employed, then the necessary SNR can be met without such a large increase in chip size and power consumption

7. THE BENEFITS OF COMMANDING

For a generic signal processor, the output is comprised of three components: signal, noise and distortion. If the processor is linear, then Figure 3-1 shows the relationship between the input and the output components as the input level increases. The output signal is proportional to the input level while noise is generally independent of the input. At high input levels, the linearity's of the system can no longer be maintained, and distortion occurs

8. EXPONENTIAL COMPANDING TECHNIQUE

We propose in this section a new nonlinear commanding technique, namely "exponential commanding", that can effectively reduce the PAPR of transmitted (commanded) OFDM signals by transforming the statistics of the amplitudes of these signals into uniform distribution. The new scheme also has the advantage of maintaining a constant average power level in the nonlinear commanding operation. The strict linearity requirements on HPA can then be partially relieved. Let, t_n^d

the d^{th} power of the amplitude of commanded signal t_n , have a uniform distribution in the interval $[0, \alpha]$ the exponent is called the degree of a specific exponential commanding scheme. The CDF t_n^d of is simply

$$F_{|t_n|^d}(x) = \frac{x}{\alpha}, \quad 0 \leq x \leq \alpha.$$

The amplitude $|t_n|$ of commanded signal has the following CDF

$$\begin{aligned} F_{|t_n|}(x) &= \text{Prob} \{ |t_n| \leq x \} \\ &= \text{Prob} \{ |t_n|^d \leq x^d \} \\ &= \frac{x^d}{\alpha}, \quad 0 \leq x \leq \sqrt[d]{\alpha}. \end{aligned}$$

The inverse function of $F_{|t_n|}(x)$ is therefore

$$F_{|t_n|}^{-1}(x) = \sqrt[d]{\alpha x}, \quad 0 \leq x \leq 1.$$

$$\begin{aligned} F_{s_n}(x) &= \text{Prob} \{ |s_n| \leq x \} \\ &= \text{Prob} \{ h(|s_n|) \leq h(x) \} \\ &= F_{|t_n|}(h(x)), \quad 0 \leq x \leq h^{-1}(\sqrt[d]{\alpha}). \end{aligned}$$

Considering the phase of input signals, the commanding function $h(x)$

$$\begin{aligned} h(x) &= \text{sgn}(x) F_{|t_n|}^{-1}(F_{|s_n|}(x)) \\ &= \text{sgn}(x) \sqrt[d]{\alpha \left[1 - \exp\left(-\frac{x^2}{\sigma^2}\right) \right]}. \end{aligned}$$

Where $\text{sgn}(x)$ is the sign function? The positive constant α determines the average power of output signals. In order to keep the input and output signals at the same average power level, we
Let

$$\alpha = \left(\frac{E[|s_n|^2]}{E\left[\sqrt[d]{1 - \exp\left(-\frac{|s_n|^2}{\sigma^2}\right)}\right]^2} \right)^{\frac{d}{2}}$$

At the receiver side, the inverse function of $h(x)$ is used in the de-commanding operation, i.e.

$$h^{-1}(x) = \text{sgn}(x) \sqrt{-\sigma^2 \log_e \left(1 - \frac{x^d}{\alpha} \right)}.$$

Fig. 2 shows the exponential commanding function $h(x)$

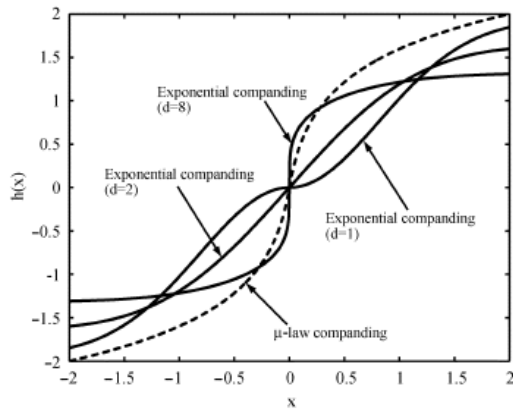


Fig. 2. The exponential companding function $h(x)$.

With degree as a parameter. Referring to (9), the commanded signals have uniformly distributed amplitudes and powers, respectively for the cases $d = 1$ and $d = 2$. When, $d \geq 2$, the proposed function $h(x)$ can compress large input signals and expand small signals simultaneously. While the μ -law commanding scheme can only enlarge small signals and does not change the signal peaks, which leads to a higher average power level of output signals. As seen, the differences between exponential commanding functions are ignorable when $d \geq 8$.

9. NEW COMPANDING ALGORITHM

OBI is the spectral leakage into alien channels. Quantification of the OBI caused by commanding requires the knowledge of the power spectral density (PSD) of the commanded signal. Unfortunately analytical expression of the PSD is in general mathematically intractable, because of

the nonlinear commanding transform involved. Here we take an alternative approach to estimate the OBI. Let (x) be a nonlinear commanding function, and $(t) = \sin(\omega t)$ be the input to the commander. The commanded signal (t) is: $(t) = [(t)] = f[\sin(\omega t)]$. Since (t) is a periodic function with the same period as (t) , (t) can then be expanded into the following Fourier series:

$$y(t) = \sum_{k=-\infty}^{+\infty} c(k)e^{jk\omega t},$$

Where the coefficients

$c(k)$ is calculated as:

$$c(k) = c(-k) = \frac{1}{T} \int_0^T y(t)e^{-jk\omega t} dt \quad T = \frac{2\pi}{\omega}.$$

Notice that the input x in this case is a pure sinusoidal signal, any $(k) \neq 0$ for $|k| > 1$ is the OBI produced by the nonlinear commanding process. Therefore, to minimize the OBI, (k) must approach to zero fast enough as k increases. It has been shown that $(k) \cdot k^{-(m+1)}$ tends to zero if $y(t)$ and its derivative up to the m -th order are continuous, or in other words, $c(k)$ converges at the rate of $k^{-(m+1)}$. Given an arbitrary number n , the n -th order derivative of $y(t)$, $dn y/dt^n$, is a function of $df(x)/dx^i$, $(i = 1, 2, \dots, n)$, as well as $\sin(\omega t)$ and $\cos(\omega t)$, i.e.:

$$\frac{d^n y}{dt^n} = g\left(\frac{d^n f(x)}{dx^n}, \frac{d^{n-1} f(x)}{dx^{n-1}}, \dots, \frac{df(x)}{dx}, \sin(\omega t), \cos(\omega t)\right)$$

Sin (ωt) and coos (ωt) are continuous functions, dny/dtn is continuous if and only if $dif(x)/dxi$ ($i = 1, 2, \dots, n$) are continuous. Based on this observation we can conclude:

Companding introduces minimum amount of OBI if the commanding function (x) is infinitely differentiable. The functions that meet the above condition are the smooth functions. We now propose a new commanding algorithm using a smooth function, namely the airy special function. The commanding function is as follows:

$$f(x) = \beta \cdot \text{sign}(x) \cdot [\text{airy}(0) - \text{airy}(\alpha \cdot |x|)],$$

Where airy (\cdot) is the airy function of the first kind. α is the parameter that controls the degree of commanding (and ultimately PAPR). β is the factor adjusting the average output power of the commander to the same level as the average input power:

$$\beta = \sqrt{\frac{E[|x|^2]}{E[|\text{airy}(0) - \text{airy}(\alpha \cdot |x|)|^2]}}$$

Where $E[\cdot]$ denotes the expectation. The decomanding function is the inverse of (x):

$$f^{-1}(x) = \frac{1}{\alpha} \cdot \text{sign}(x) \cdot \text{airy}^{-1} \left[\text{airy}(0) - \frac{|x|}{\beta} \right],$$

Where the superscript-1 represents the inverse operation. Notice that the input to the deco pander is a quantized signal with finite set of values. We can therefore numerically pre-compute $f^{-1}(x)$ and use table look-up to perform the decomanding in practice. Next we examine the BER performance of the algorithm. Let (t) denote the output signal of the commander, (t) the white Gaussian noise. The received signal can be expressed as:

$$z(t) = y(t) + w(t). \quad \text{The decomanded signal } \tilde{x}(t) \text{ simply is:}$$

$$\tilde{x}(t) = f^{-1}[z(t)] = f^{-1}[y(t) + w(t)].$$

10. RESULTS:

Notice that the signal-to-noise ratio (SNR) in a typical additive white Gaussian noise (AWGN) Channel is much greater than



Fig:Performance curve of ofdm

Using the first order Taylor series expansion,

$$\tilde{x}(t) \approx x(t) + \frac{df^{-1}(u)}{du} \Big|_{u=y(t)} \cdot w(t).$$

From the given Equation shows that if (t) falls into the range of the decompanding function $f^{-1}(u)$ where $df^{-1}(u)/du|_{u=y(t)} < 1$, the noise $w(t)$ is suppressed, and if $y(t)$ is out of the range, $df^{-1}(u)/du|_{u=y(t)} > 1$ and the noise is enhanced. Therefore, if the parameter α in (8) is properly chosen such that more (t) is within the noise-suppression range of $f^{-1}(u)$, it is possible to achieve better overall BER performance. It is worth to mention though that BER and PAPR affect each other adversely and therefore there is a tradeoff to make.

CONCLUSION:

In this paper, we have proposed a new commanding algorithm. Both theoretical analysis and computer simulation show that the algorithm

offers improved performance in terms of BER and OBI while reducing PAPR effectively.

REFERENCES

- [1] R. van Nee and R. Prasad, *OFDM for Wireless Multimedia Communications*. Boston, MA: Barteck House, 2000.
- [2] S. H. Han and J. H. Lee, "An Overview of peak-to-average power ratio reduction techniques for multicarrier transmission," *IEEE Wireless Commun.*, vol. 12, pp. 56-65, Apr. 2005.
- [3] X. Wang, T. T. Thong, and C. S. Ng, "Reduction of peak-to-average power ratio of OFDM system using a commanding technique," *IEEE Trans. Broadcast.*, vol. 45, no. 3, pp. 303-307, Sept. 1999.
- [4] T. Jiang and G. Zhu, "Nonlinear companding transform for reducing peak to- average power ratio of OFDM signals," *IEEE Trans. Broadcast.*, vol. 50, no. 3, pp. 342-346, Sept. 2004.
- [5] T. Jiang, Y. Yang, and Y. Song, "Exponential commanding technique for PAPR reduction in OFDM systems," *IEEE Trans. Broadcast.*, vol. 51, no. 2, pp. 244-248, June 2005.
- [6] D. Lowe and X. Huang, "Optimal adaptive hyperbolic commanding for OFDM," in *Proc. IEEE Second Intl Conf. Wireless Broadband and Ultra Wideband Common.*, pp. 24-29, Aug. 2004.
- [7] J. Song and H. Ochiai, "FPGA implementation of peak cancellation for PAPR reduction of

- OFDM signals,” in *Proc. IEEE Int. Conf. Commun. Syst. (ICCS'14)*, pp. 414–418, Nov. 2014.
- [8] J. Song and H. Ochiai, “A low-complexity peak cancellation scheme and its FPGA implementation for peak-to-average power ratio reduction,” *EURASIP J. Wireless Commun. Netw.*, vol. 2015, pp. 1–14, Mar. 2015.
- [9] H. B. Jeon, J. S. No, and D. J. Shin, “A new PAPR reduction scheme using efficient peak cancellation for OFDM systems,” *IEEE Trans. Broadcast.*, vol. 58, no. 4, pp. 619–628, Dec. 2012.
- [10] L. Wang and C. Tellambura, “Analysis of clipping noise and Tone-Reservation algorithms for peak reduction in OFDM systems,” *IEEE Trans. Veh. Technol.*, vol. 57, no. 3, pp. 1675–1694, May 2008.
- [11] H. Ochiai and H. Imai, “Performance of the deliberate clipping with adaptive symbol selection for strictly band-limited OFDM systems,” *IEEE J. Sel. Areas Commun.*, vol. 18, no. 11, pp. 2270–2277, Nov. 2000.
- [12] R. Dinis and A. Gusmao, “A class of nonlinear signal-processing schemes for bandwidth-efficient OFDM transmission with low envelope fluctuation,” *IEEE Trans. Commun.*, vol. 52, no. 10, pp. 2009–2018, Oct. 2004.
- [13] D. Dardari, V. Tralli, and A. Vaccari, “A theoretical characterization of nonlinear distortion effects in OFDM systems,” *IEEE Trans. Commun.*, vol. 48, no. 10, pp. 1755–1764, Oct. 2000.
- [14] E. Costa and S. Pupolin, “M-QAM-OFDM system performance in the presence of a nonlinear amplifier and phase noise,” *IEEE Trans. Commun.*, vol. 50, no. 3, pp. 462–472, Mar. 2002.
- [15] Y. Xiao, W. Bai, L. Dan, G. Wu, and S. Li, “Performance analysis of peak cancellation in OFDM systems,” *Sci. China Inf. Sci.*, vol. 55, pp. 789–794, Apr. 2012.
- [16] R. Price, “A useful theorem for nonlinear devices having Gaussian inputs,” *IRE Trans. Inf. Theory*, vol. 4, no. 2, pp. 69–72, Jun. 1958.
- [17] J. J. Bussgang, “Crosscorrelation functions of amplitude-distorted Gaussian signals,” *Res. Lab. Electron., Massachusetts Inst. Technol.*, Cambridge, MA, USA, Tech. Rep. 216, Mar. 1952.
- [18] H. Rowe, “Memoryless nonlinearities with Gaussian inputs: Elementary results,” *Bell Syst. Tech. J.*, vol. 61, pp. 1519–1526, Sep. 1982.
- [19] J. G. Proakis and M. Salehi, *Digital Communications*, 5th ed. New York, NY, USA: McGraw-Hill, 2008.
- [20] H. Ochiai, “An analysis of band-limited communication systems from amplifier efficiency and distortion perspective,” *IEEE Trans.*



Commun., vol. 61, no. 4, pp. 1460–1472, Feb. 2013.

[21] P. Banelli and S. Cacopardi, “Theoretical analysis and performance of OFDM signals in nonlinear AWGN channels,” *IEEE Trans. Commun.*, vol. 48, no. 3, pp. 430–441, Mar. 2000.

[22] A. Y. Kibangou and G. Favier, “Wiener-Hammerstein systems modeling using diagonal Volterra kernels coefficients,” *IEEE Signal Process. Lett.*, vol. 13, no. 6, p. 381, Jun. 2006.

[23] S. O. Rice, “Mathematical analysis of random noise,” *Bell Syst. Tech. J.*, vol. 23, pp. 282–332, Jul. 1944.

[24] A. R. Bahai, M. Singh, A. J. Goldsmith, and B. R. Saltzberg, “A new approach for evaluating clipping distortion in multicarrier systems,” *IEEE J. Sel. Areas Commun.*, vol. 20, no. 5, pp. 1037–1046, Jun. 2002.